

# Control Integrated POver System (CIPOS™)

IFCM20U65GD

Datasheet

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CIPOS™

## Control Integrated POver System

*Dual In-Line Intelligent Power Module*

*Three Phase Interleaved PFC 650V / 20A*

### Features

Dual In-Line molded module

- TRENCHSTOP™ 5 IGBTs
- Rapid switching emitter controlled diode
- Rugged SOI gate driver technology with stability against transient
- Over current shutdown
- Under-voltage lockout
- All of 3 switches turn off during protection
- Temperature monitor
- Emitter pins accessible for all phase current monitoring (open emitter)
- Lead-free terminal plating; RoHS compliant
- Very low thermal resistance due to DCB

### Target Applications

- 3-Phase Interleaved PFC

### Description

The CIPOS™ module family offers the chance for integrating various power and control components to increase reliability, optimize PCB size and system costs.

It is designed to enhance the system efficiency by improvement of power factor. The package concept is specially adapted to power applications, which need good thermal conduction and electrical isolation, but also EMI-save control and overload protection.

TRENCHSTOP™ 5 IGBTs are combined with an optimized SOI gate driver for excellent electrical performance.

### System Configuration

- 3-Phase Interleaved PFC with TRENCHSTOP™ 5 IGBTs and Rapid switching emitter controlled diode
- SOI gate driver
- Thermistor
- Pin-to-heatsink clearance distance typ. 1.6mm

## Pin Configuration

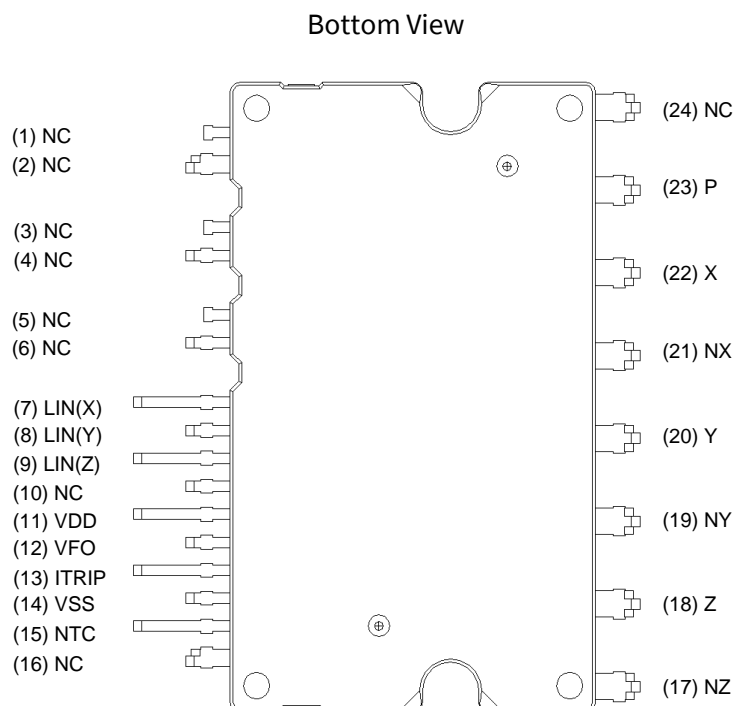


Figure 1 Pin configuration

## Internal Electrical Schematic

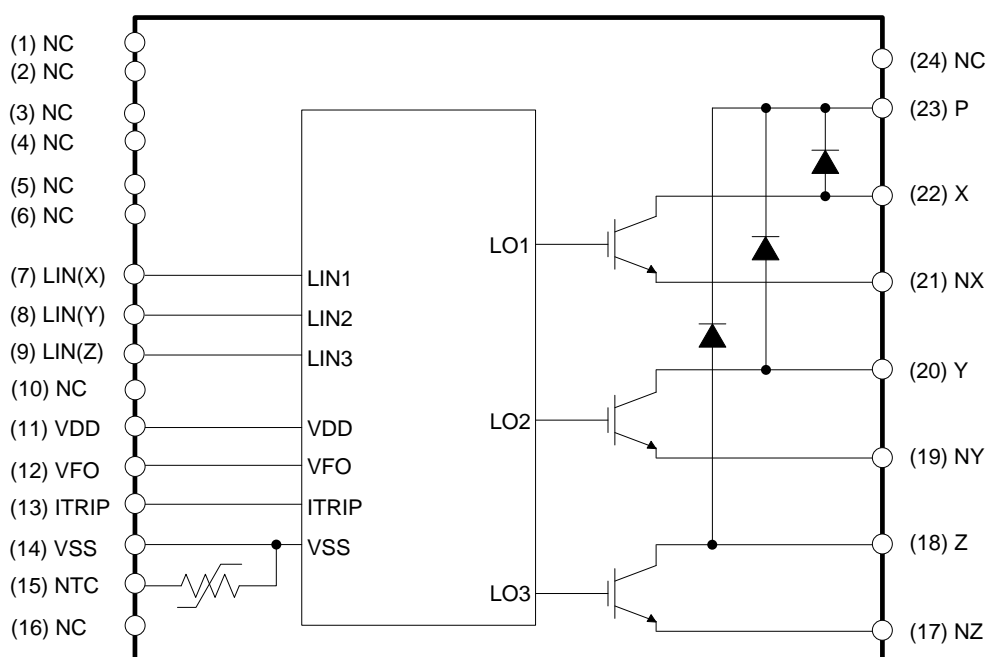


Figure 2 Internal schematic

## Pin Assignment

Pin Number	Pin Name	Pin Description
1	NC	No Connection
2	NC	No Connection
3	NC	No Connection
4	NC	No Connection
5	NC	No Connection
6	NC	No Connection
7	LIN(X)	X phase IGBT gate driver input
8	LIN(Y)	Y phase IGBT gate driver input
9	LIN(Z)	Z phase IGBT gate driver input
10	NC	No Connection
11	VDD	Control supply
12	VFO	Fault output
13	ITRIP	Over current shutdown input
14	VSS	Control negative supply
15	NTC	Thermistor
16	NC	No Connection
17	NZ	Z phase IGBT emitter
18	Z	Z phase IGBT collector
19	NY	Y phase IGBT emitter
20	Y	Y phase IGBT collector
21	NX	X phase IGBT emitter
22	X	X phase IGBT collector
23	P	Positive output voltage
24	NC	No Connection

## Pin Description

### LIN(X, Y, Z) (IGBT control pins, Pin 7, 8, 9)

These pins are positive logic and they are responsible for the control of the integrated IGBT. The Schmitt-trigger input thresholds of them are such to guarantee LSTTL and CMOS compatibility down to 3.3V controller outputs. Pull-down resistor of about 5k $\Omega$  is internally provided to pre-bias inputs during supply start-up and a zener clamp is provided for pin protection purposes. Input Schmitt-trigger and noise filter provide beneficial noise rejection to short input pulses.

The noise filter suppresses control pulses which are below the filter time  $t_{FILIN}$ . The filter acts according to Figure 4.

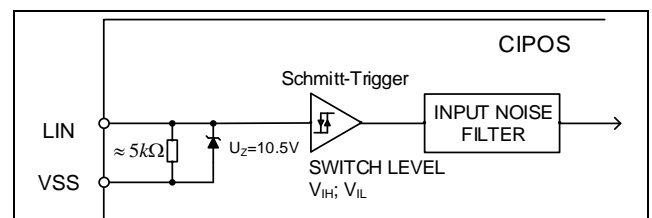


Figure 3 Input pin structure

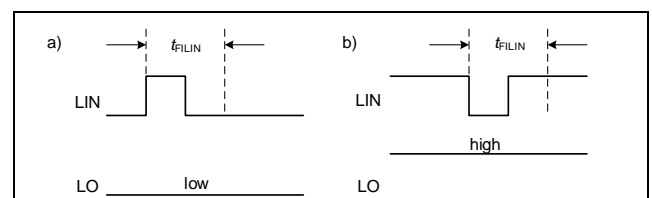


Figure 4 Input filter timing diagram

It is recommended for proper work of this product not to provide input pulse-width lower than 0.5us.

#### VFO (Fault-output, Pin 12)

The VFO pin indicates a module failure in case of under voltage at pin VDD or in case of triggered over current detection at ITRIP.

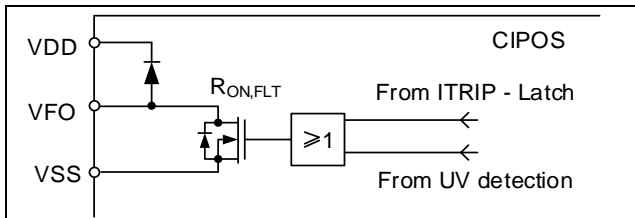


Figure 5 Internal circuit at pin VFO

#### NTC (Thermistor, Pin 15)

The NTC pin provides direct access to thermistor, which is referenced to VSS. An external pull-up resistor connected to +5V ensures that the resulting voltage can be directly connected to the microcontroller.

#### ITRIP (Over current detection function, Pin 13)

CIPOS™ provides an over current detection function by connecting the ITRIP input with the IGBT collector current feedback. The ITRIP comparator threshold (typ. 0.47V) is referenced to VSS ground. An input noise filter (typ.:  $t_{ITRIPMIN} = 530ns$ ) prevents the driver to detect false over-current events.

Over current detection generates a shutdown of all outputs of the gate driver after the shutdown propagation delay of typically 1000ns.

#### VDD, VSS (Control supply and reference, Pin 11, 14)

VDD is the control supply and it provides power both to input logic and to output power stage. Input logic is referenced to VSS ground.

The under-voltage circuit enables the device to operate at power on when a supply voltage of at least a typical voltage of  $VDD_{UV+} = 12.1V$  is present.

The IC shuts down all the gate drivers power outputs, when the VDD supply voltage is below  $VDD_{UV-} = 10.4V$ . This prevents the external power switches from critically low gate voltage levels during on-state and therefore from excessive power dissipation.

#### NX, NY, NZ (IGBT emitter, Pin 17, 19, 21)

The IGBT emitters are available for current measurements of each phase. It is recommended to keep the connection to pin VSS as short as possible in order to avoid unnecessary inductive voltage drops.

#### X, Y, Z (IGBT collector, Pin 18, 20, 22)

These pins are IGBT collector. It is mandatory to connect anti-parallel diode between IGBT collector and emitter.

#### P (Positive output voltage, Pin 23)

The diode cathodes are connected to the output voltage. It is noted that the voltage does not exceed 450 V.

## Absolute Maximum Ratings

( $V_{DD} = 15V$  and  $T_j = 25^\circ C$ , if not stated otherwise)

### Module Section

Description	Condition	Symbol	Value		Unit
			min	max	
Storage temperature range		$T_{stg}$	-40	125	$^\circ C$
Isolation test voltage	RMS, $f=60Hz$ , $t=1min$	$V_{ISOL}$	2000	-	V
Operating case temperature range	Refer to Figure 6	$T_C$	-40	100	$^\circ C$

### Power Section

Description	Condition	Symbol	Value		Unit
			min	max	
DC link output voltage of P-N	Applied between P-N	$V_{PN}$	-	450	V
DC link output voltage (surge) of P-N	Applied between P-N	$V_{PN(surge)}$	-	500	V
Max. blocking voltage	$I_C = 250\mu A$	$V_{CES}$	650	-	V
Repetitive peak reverse voltage	$I_R = 250\mu A$	$V_{RRM}$	650	-	V
Input RMS current of each phase	$T_J \leq 150^\circ C$ , $T_C = 25^\circ C$ $T_C = 80^\circ C$	$I_i$	-	20 15	A
Maximum peak input current of each phase	$T_J \leq 150^\circ C$ , $T_C = 25^\circ C$ less than 1ms, non-repetitive	$I_{i(peak)}$	-	60	A
Power dissipation of each IGBT		$P_{tot}$	-	52.3	W
Operating junction temperature range		$T_J$	-40	150	$^\circ C$
Single IGBT thermal resistance, junction-case		$R_{thJC}$	-	2.39	K/W
Single diode thermal resistance, junction-case		$R_{thJCD}$	-	2.77	K/W

## Control Section

Description	Condition	Symbol	Value		Unit
			min	max	
Module supply voltage		$V_{DD}$	-1	20	V
Input voltage	LIN, ITRIP	$V_{IN}$ $V_{ITRIP}$	-1 -1	10	V
Switching frequency		$f_{PWM}$	-	60	kHz

## Recommended Operation Conditions

All voltages are absolute voltages referenced to  $V_{SS}$  -potential unless otherwise specified.

Description	Symbol	Value			Unit
		min	typ	max	
DC link output voltage of P-N	$V_{PN}$	0	-	450	V
Control supply voltage	$V_{DD}$	13.5	15	16.5	V
Control supply variation	$\Delta V_{DD}$	-1	-	1	V/ $\mu$ s
Logic input voltages LIN, ITRIP	$V_{IN}$	0	-	5	V
	$V_{ITRIP}$	0	-	5	V
Between $V_{SS}$ - N (including surge)	$V_{SS}$	-5	-	5	V

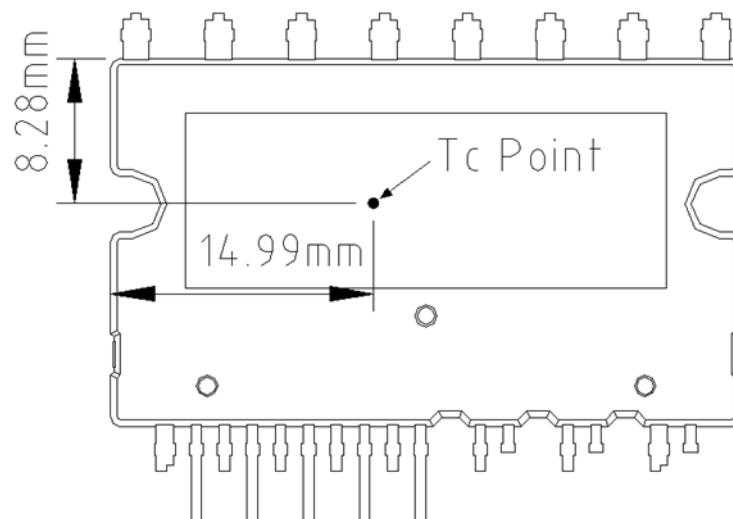


Figure 6  $T_c$  measurement point<sup>1</sup>

<sup>1</sup>Any measurement except for the specified point in figure 6 is not relevant for the temperature verification and brings wrong or different information.



## Static Parameters

( $V_{DD} = 15V$  and  $T_j = 25^\circ C$ , if not stated otherwise)

Description	Condition	Symbol	Value			Unit
			min	typ	max	
Collector-Emitter saturation voltage	$I_C = 15A$ $T_j = 25^\circ C$ $150^\circ C$	$V_{CE(sat)}$	- -	1.7 1.9	2.3 -	V
Diode forward voltage	$I_F = 15A$ $T_j = 25^\circ C$ $150^\circ C$	$V_F$	- -	1.4 1.3	1.85 -	V
Collector-Emitter leakage current	$V_{CE} = 650V$	$I_{CES}$	-	-	1	mA
Diode reverse leakage current	$V_R = 650V$	$I_R$	-	-	1	mA
Logic "1" input voltage (LIN)		$V_{IH}$	-	2.1	2.5	V
Logic "0" input voltage (LIN)		$V_{IL}$	0.7	0.9	-	V
ITRIP positive going threshold		$V_{IT,TH+}$	400	470	540	mV
ITRIP input hysteresis		$V_{IT,HYS}$	40	70	-	mV
VDD supply under voltage positive going threshold		$VDD_{UV+}$	10.8	12.1	13.0	V
VDD supply under voltage negative going threshold		$VDD_{UV-}$	9.5	10.4	11.2	V
VDD supply under voltage lockout hysteresis		$VDD_{UVH}$	1.0	1.7	-	V
Quiescent VDD supply current	$L_{IN} = 0V$	$I_{QDD}$	-	370	900	$\mu A$
Input bias current	$V_{IN} = 5V$	$I_{IN+}$	-	1	1.5	mA
Input bias current	$V_{IN} = 0V$	$I_{IN-}$	-	2	-	$\mu A$
ITRIP input bias current	$V_{ITRIP} = 5V$	$I_{ITRIP+}$	-	65	150	$\mu A$
VFO input bias current	$VFO = 5V, V_{ITRIP} = 0V$	$I_{FO}$	-	2	-	nA
VFO output voltage	$I_{FO} = 10mA, V_{ITRIP} = 1V$	$V_{FO}$	-	0.5	-	V

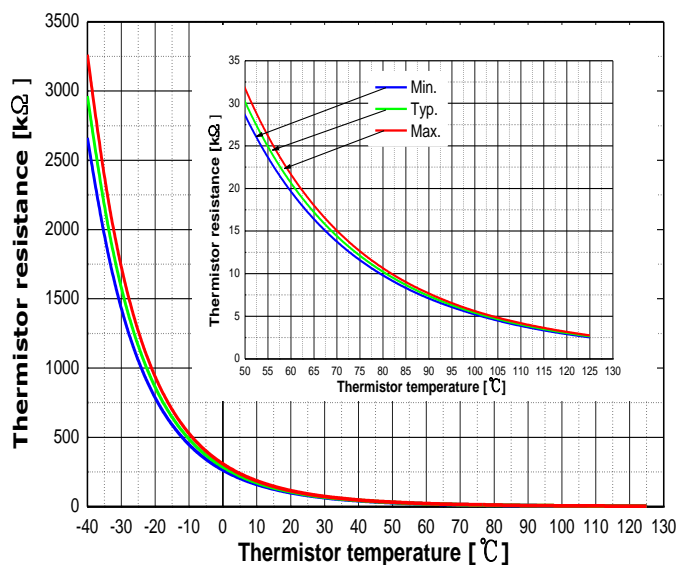
## Dynamic Parameters

( $V_{DD} = 15V$  and  $T_J = 25^\circ C$ , if not stated otherwise)

Description	Condition	Symbol	Value			Unit
			min	typ	max	
Turn-on propagation delay time	$V_{LIN} = 5V$ , $I_C = 15A$ , $V_{DC} = 400V$	$t_{on}$	-	585	-	ns
Turn-on rise time		$t_r$	-	20	-	ns
Turn-on switching time		$t_{c(on)}$	-	105	-	ns
Reverse recovery time		$t_{rr}$	-	100	-	ns
Turn-off propagation delay time	$V_{LIN} = 0V$ , $I_C = 15A$ , $V_{DC} = 400V$	$t_{off}$	-	630	-	ns
Turn-off fall time		$t_f$	-	10	-	ns
Turn-off switching time		$t_{c(off)}$	-	20	-	ns
Input filter time ITRIP	$V_{ITRIP} = 1V$	$t_{ITRIPmin}$	-	530	-	ns
Input filter time at LIN for turn on and off	$V_{LIN} = 0V \text{ \& } 5V$	$t_{FILIN}$	-	290	-	ns
Fault clear time after ITRIP-fault	$V_{ITRIP} = 1V$	$t_{FLTCLR}$	40	-	-	$\mu s$
IGBT turn-on energy (includes reverse recovery of diode)	$V_{DC} = 400V$ , $I_C = 15A$ $T_J = 25^\circ C$ $150^\circ C$	$E_{on}$	-	440	-	$\mu J$
			-	550	-	
IGBT turn-off energy	$V_{DC} = 400V$ , $I_C = 15A$ $T_J = 25^\circ C$ $150^\circ C$	$E_{off}$	-	35	-	$\mu J$
			-	65	-	
Diode recovery energy	$V_{DC} = 400V$ , $I_C = 15A$ $T_J = 25^\circ C$ $150^\circ C$	$E_{rec}$	-	75	-	$\mu J$
			-	145	-	

## Thermistor

Description	Condition	Symbol	Value			Unit
			min	typ	max	
Resistor	$T_{NTC} = 25^{\circ}\text{C}$	$R_{NTC}$	-	85	-	$k\Omega$
B-constant of NTC (Negative Temperature Coefficient)		$B(25/100)$	-	4092	-	K



T [°C]	Rmin. [kΩ]	Rtyp. [kΩ]	Rmax. [kΩ]
50	28.400	29.972	31.545
60	19.517	20.515	21.514
70	13.670	14.315	14.960
80	9.745	10.169	10.593
90	7.062	7.345	7.628
100	5.199	5.388	5.576
110	3.856	4.009	4.163
120	2.900	3.024	3.149
125	2.527	2.639	2.751

Figure 7 Thermistor resistance – temperature curve and table

(For more information, please refer to the application note 'AN CIPOS™-Mini 1 Technical description')

## Mechanical Characteristics and Ratings

Description	Condition	Value			Unit
		min	typ	max	
Mounting torque	M3 screw and washer	0.49	-	0.78	Nm
Flatness	Refer to Figure 8	-50	-	100	$\mu\text{m}$
Weight		-	6.58	-	g

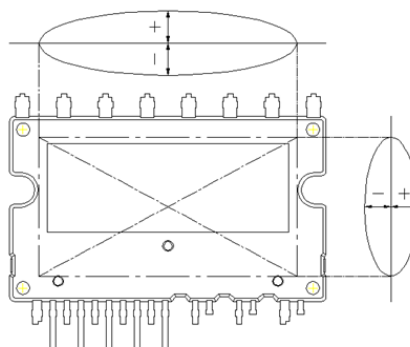
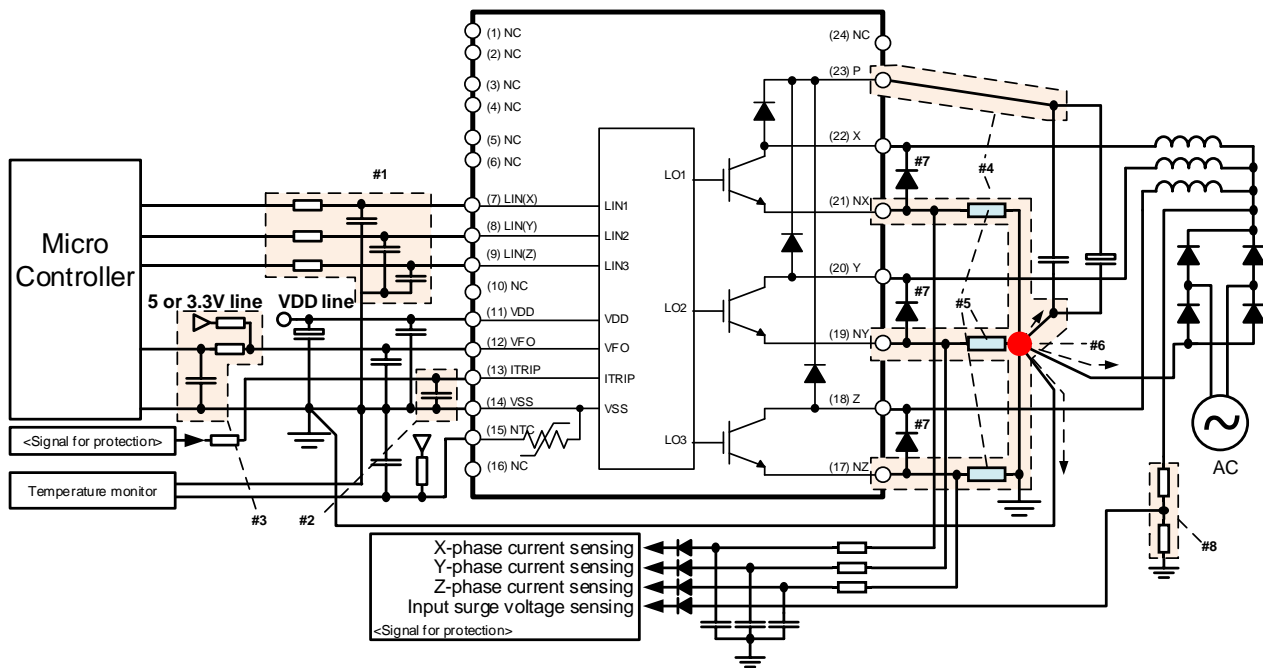


Figure 8 Flatness measurement position

## Circuit of a Typical Application



**Figure 9** Typical application circuit

Because CIPOS™ Mini PFC has very high speed switching characteristics, considerable large surge voltage between P and N terminals and switching noise on signaling path are generated easily. Please pay attention to the below items for optimized application circuit design.

### 1. Input circuit

- To reduce input signal noise by high speed switching, the  $R_{IN}$  and  $C_{IN}$  filter circuit should be mounted. (100Ω, 1nF)
- $C_{IN}$  should be placed as close to  $V_{SS}$  pin as possible.

### 2. Itrip circuit

- To prevent protection function errors,  $C_{ITRIP}$  should be placed as close to Itrip and  $V_{SS}$  pins as possible.

### 3. VFO circuit

- VFO output is an open drain output. This signal line should be pulled up to the positive side of the 5V/3.3V logic power supply with a proper resistor  $R_{PU}$ . It is recommended that RC filter be placed as close to the controller as possible.

### 4. Snubber capacitor

- The wiring between CIPOS™ Mini PFC and snubber capacitor including shunt resistor should be as short as possible.

### 5. Shunt resistor

- The shunt resistor of SMD type should be used for reducing its stray inductance.

### 6. Ground pattern

- Ground pattern should be separated at only one point of shunt resistor as short as possible.

### 7. It is mandatory to connect anti-parallel diode (2A, voltage rating higher than 650V) to PFC IGBT.

### 8. Input surge voltage protection circuit

- This protection circuit is necessary for PFC IGBT to be protected from excessive surge voltage.

## Switching Times Definition

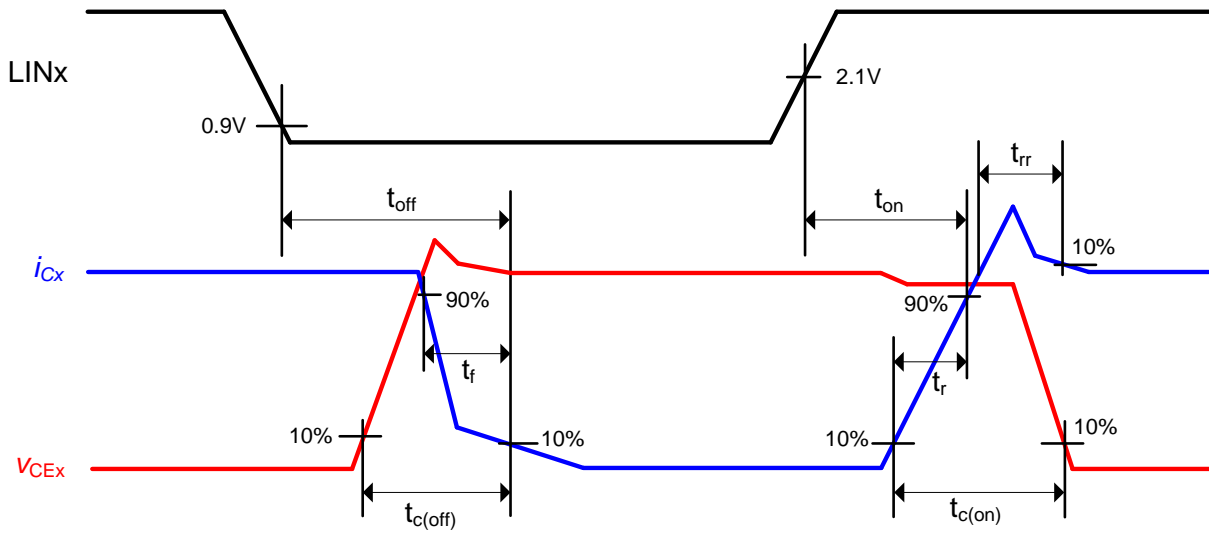
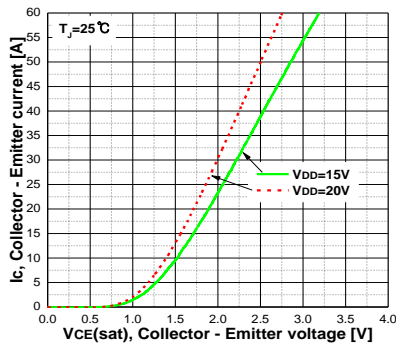
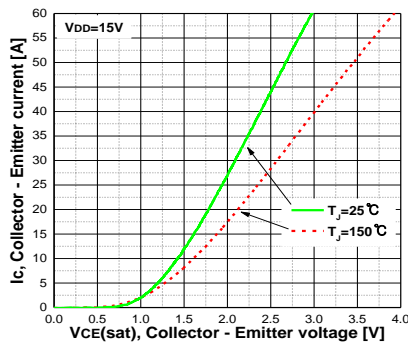


Figure 10 Switching times definition

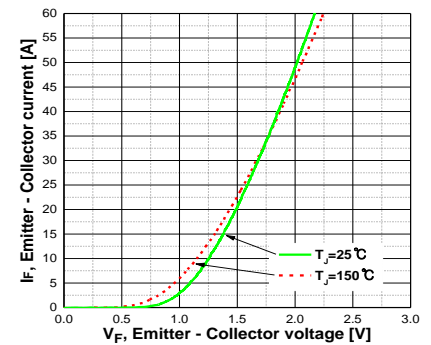
## Electrical characteristic



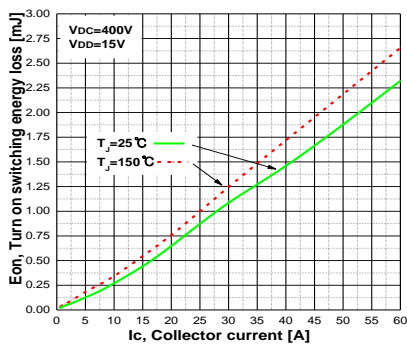
Typ. Collector - Emitter saturation voltage



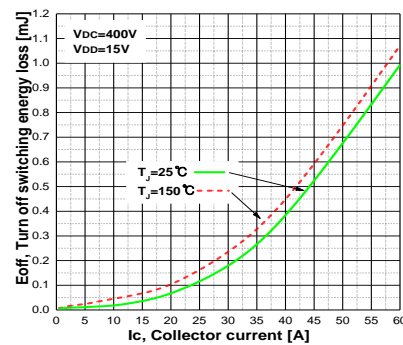
Typ. Collector - Emitter saturation voltage



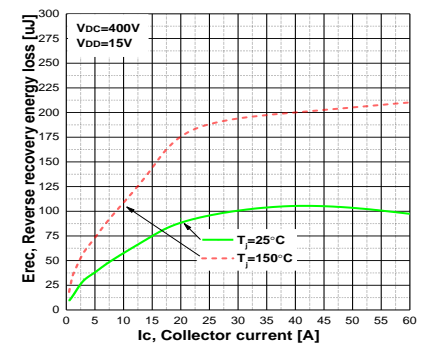
Typ. Emitter - Collector forward voltage



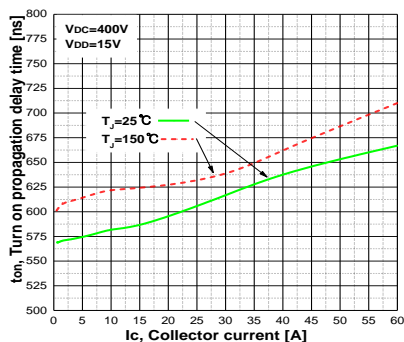
Typ. Turn on switching energy loss



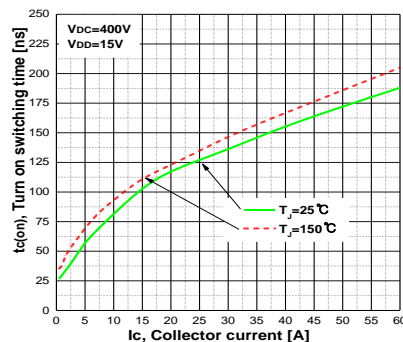
Typ. Turn off switching energy loss



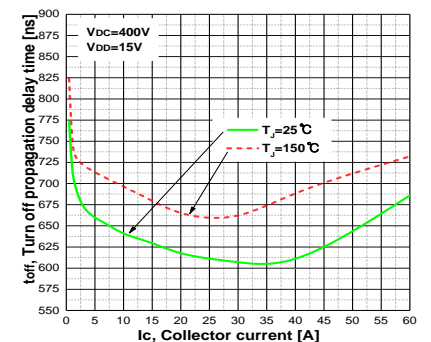
Typ. Reverse recovery energy loss



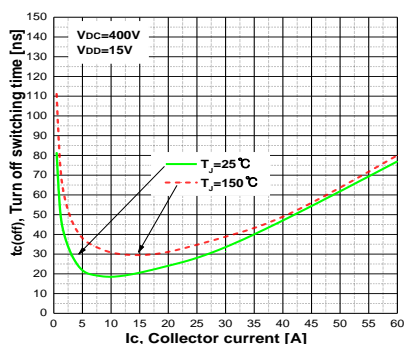
Typ. Turn on propagation delay time



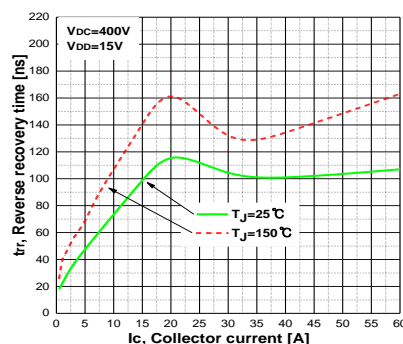
Typ. Turn on switching time



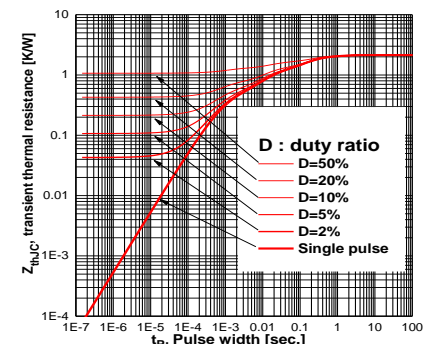
Typ. Turn off propagation delay time



Typ. Turn off switching time

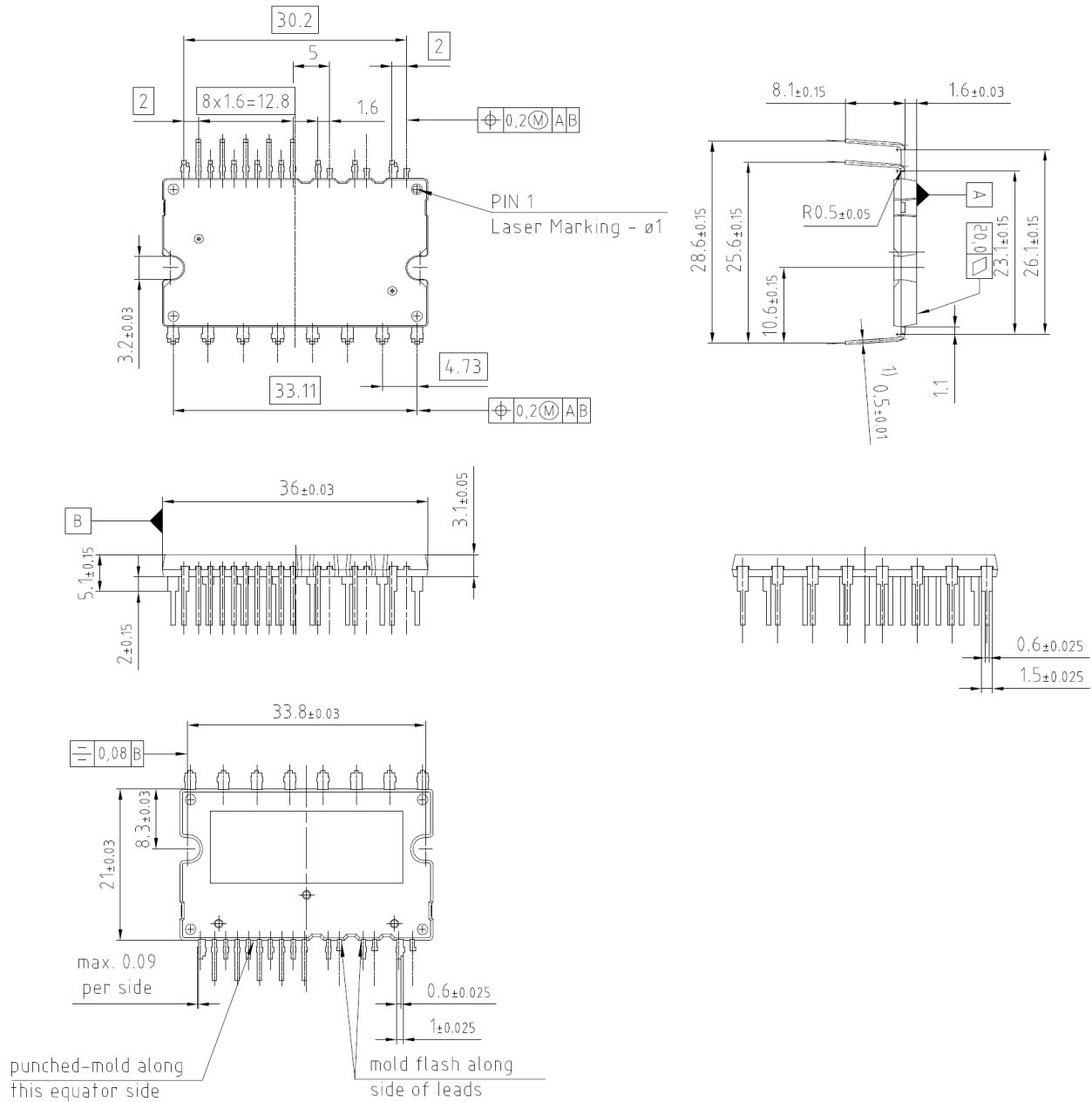


Typ. Reverse recovery time



IGBT transient thermal resistance  
at all IGBTs operation

## Package Outline



## Revision History

Major changes since the last revision

Page or Reference	Description of change



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